Assessing the Impact of Reactive Power Droop on Inverter Based Microgrid Stability

Ahmed Lasheen, Mohammed Ammar, Hatem Zeineldin, Ahmed Al-Durra, Ehab El-Saadany and Mostafa F. Shaaban

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2 Abstract—Droop control is the most common approach fo#7 3 controlling inverter-based micro-grids. The active power droops gain has always been considered as the main parameter forg 4 identifying the micro-grid stability margin. Increasing this margin 05 6 improves the transient performance and provides robustness to the micro-grid for a wide range of operations. Previous work on $\overline{1}^1$ 7 8 droop control focused on the active power droop gain, which is² 9 required for accurate power sharing as well as for micro-grid3 10 stability assessment. This paper utilizes small-signal stability4 analysis to analyze the impact of the reactive power droop gain on 5 micro-grid stability, which is ignored in previous work 5_6 Consequently, a micro-grid domain of stability chart is proposed 12 13 and defined in the mp_{max} -nq plane, which represents the zone⁷ 14 within which the micro-grid will maintain stable operation. The δ^8 15 16 proposed domain of stability chart is utilized to assess and9 17 compare the impact of the conventional and proportionat derivative (PD) reactive power droop controller on the micro-grid 1 18 stability margin. The results show that there exists a reactive $\frac{1}{2}$ 19 20 power droop gain at which the stability margin is minimum. Furthermore, it has been shown, through the domain of stability³ 21 22 chart, that the PD reactive power droop controller is capable an@4 23 sufficient to significantly increase the micro-grid stability margin5 24 while maintaining equal load sharing. Further, the domain of 6 25 stability chart can serve as a useful tool for defining the micro-grid $_7$ droop gain operational boundaries and for assessing and 26 comparing inverter-based micro-grid control schemes. 69

27 28 29 Index Terms—Reactive power droop, stability margin, domain₀ of stability, inverter-based micro-grid, and small-signal stability. $_{71}^{70}$ 30

I. INTRODUCTION

73 The utilization of renewable energy resources, especiall y_4 32 33 wind and solar energy, is growing rapidly to not only meets the massive increase in demand but also to reduce greenhous96 34 35 gas emissions and increase energy security by diversifying the7 power generation sources. Renewable energy resources coupled 36 37 with energy storage are among the main components, yet $als \varphi_{Q}$ drivers for the growing interest in micro-grids. IEEE Std₈₀ 38 1547.4 provides guidelines and recommendations for operating 39 40 and designing microgrids and necessitates the development of 2tools for assessing its operation [1]. Microgrids are defined as₃ 41 a set of Distributed Generation (DG) connected to supply and d_A 42 set of loads [2]- [3]. A microgrid can be operated in either 5 43 islanded or grid-connected modes [4]. In an islanded mode o_{66} 44

45 operation, each DG should be capable of providing the required

share of the total power demand based on its rating [5]. In general, microgrids are equipped with primary and secondary control where primary control focuses on stabilizing the microgrid, for a given load active and reactive power demand, at a frequency and voltage within acceptable levels. The micro-grid secondary control is responsible for restoring both the frequency and voltage to their nominal values for the given load active and reactive demand. The active and reactive power droop characteristics represent the relationship between active and reactive power with the system frequency and voltage, respectively.

According to the literature, two possible control strategies for islanded microgrids are widely adopted [6]. The first control strategy is based on a centralized controller and a communication network collecting data from local DGs. However, due to possible communication delays and communication line disturbances, the reliability of the microgrid can deteriorate [7]. The second and most common control strategy is based on the droop control technique, which does not require any communication between the DGs [8]- [9]. An exhaustive review on droop controlled microgrids is given in [10], while microgrid stability issues are discussed in [11].

Micro-grid stability impact and enhancement studies have been one of the main challenges and focus in the past few years. In [12], stability analysis based on the small-signal state-space model of an Inverter-Based DG (IBDG) in an islanded microgrid is introduced, and the dependence of the dominant low-frequency modes on the power-sharing controller and network configuration has been highlighted. In [13], an adaptive decentralized droop controller is designed to improve the transient performance of the microgrid. Both active and reactive power droop controllers are modified by adding a derivative controller. The stability of a hybrid microgrid that includes both IBDGs and a diesel generator is investigated under equal and unequal power-sharing conditions in [14]. The optimal active power droop gain of the diesel generator is obtained to improve the stability of the microgrid. In [15], a stability analysis based on the frequency domain is introduced, where the optimal values of the local DG controller are obtained using a genetic algorithm.

The stability margin for the micro-grid constraints the active

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power droop gain value, limiting the secondary control action 58 1 2 Secondary control relies on changing and optimizing the droop 9 3 gains to restore system frequency and voltage [16]- [17]. Thus 0 4 it is important to have a wider range for the active power droof gain that can achieve acceptable transient performance whil62 5 maintaining micro-grid stability. This can be achieved b\$3 6 7 increasing the stability margin, and consequently, the domai64 8 of stability. The most dominant poles for droop-based micro65 9 grids are complex, and thus in [18], the angle between the6 10 dominant eigenvalues and the imaginary axis is a sufficien67 measure for determining the stability margin. The experimentals 11 microgrid setup in [19] is characterized by a decrease of 9 12 13 stability when setting high droop gains to improve the transient() 14 response. The stability margin is improved in [20] using **a**₁ 15 supplementary droop loop while the work in [21] enhances it2 16 by adding an adaptive feed-forward compensation to eliminat@3 17 the effect of droop gains on system stability. In [22], a virtual 4 18 impedance is proposed to improve the stability margin. The5 19 sensitivity analysis in [23] shows that increasing the capacity of 6 20 one of the distributed energy sources provides higher damping7 21 and improves the microgrid's stability margin. In [24], 78 22 comparison between the stability margins of microgrid³⁷9 23 operating with conventional, generalized, and transient droogo 24 concludes that the conventional droop microgrid suffers from 1 25 the worst stability margin with the lowest maximum actives2 26 power droop gain. 83

27 Designing a power filter is proposed in [25] to increase the 28 stability with conventional droop and to mitigate the effect of 5 29 line dynamics. Improvement of the stability margin is achieves 6 30 in [26] by adding cascaded lead compensators in the active7 31 power droop controller. The analysis was performed and 32 compared when using a single lead controller, two cascade 8^8 33 lead controllers, and three cascaded lead controllers. A stat89 34 feedback of the phase angle is implemented in [27] to improv90 the stability margin. The domain of stability region, develope91 35 in [14], focused on the active power droop gains for both 2 36 37 inverter and synchronous based DGs. 93

38 Properly sized virtual impedance using Lyapunov stabilit94 39 method is proposed in [28] to manage the instability issues in 5 40 droop-controlled microgrids. Small signal stability of practica96 41 microgrids is addressed in [29], and analytical stabilit97 42 conditions that include droop gains and network parameters ar98 43 provided. In [30], the instability problem associated with those 44 low frequency power modes in droop-controlled microgrids 100 45 addressed by utilizing a robust stabilizer to damp these mode01 and consequently enhance stability and power sharing. In [31] 46 an adaptive virtual impedance and adaptive droops are adopted 47 to manage intermittent energy sources. Optimal tuning of drobb3 48 gains is proposed in [32] to enhance frequency regulation and⁴ 49 improve microgrid stability. However, the optimization⁵ 50 51 problem will be restricted to the maximum droop gain limits⁶ that result in marginal stability. Eigenvalues associated with the? 52 droop gains and stability boundaries are investigated 10^8 53 compare between different reduced-order network models 1199 54 55 [33]. In [34], the stability and robustness of the islanded 0microgrid are improved by proposing a linear quadratic¹ 56 regulator to replace conventional droop controller to overconhe2 57

the stability problems associate with the droop gains and low pass filters.

Based on the aforementioned discussion, it is clear that all the previous work [16-34] analyzed the microgrid stability without taking into consideration the impact of the reactive power droop, which is the main focus of the proposed work.

This paper first provides a comprehensive impact assessment of the reactive power droop gain on the micro-grid stability margin. A micro-grid stability domain operational region, represented in the $m_p \max -n_q$ plane, is proposed and developed using eigenvalue analysis. The domain of stability is the region which encompasses the active and reactive droop gains that result in stable micro-grid operation. Any values of droop gains outside this region will lead to instability. To maintain equal active power load sharing, the conventional active power droop controller is utilized while the PD reactive power controller is implemented to enhance the stability margin. The domain of stability region is determined to assess and quantify the enhancement in the micro-grid stability margin using the proportional derivative reactive power droop.

The paper is organized as follows. In Section II, the smallsignal model is developed to represent the microgrid with the proposed reactive power PD controller. Eigenvalue analysis for the small-signal model is introduced in Section III to determine the maximum active power gain that ensures stability. In Section IV, simulation results of a benchmark IBDG micro-grid with the droop control is presented to analyze the impacts of the droop control parameters on the domain of stability and transient performance. Finally, conclusions are discussed in Section V.

II. SMALL-SIGNAL MODEL

In this section, a small-signal model of the inverter-based microgrid considering the PD reactive power droop controller is developed. It is a modification of the small-signal model of the IBDG micro-grid, introduced in [12], used to establish the relationship between the various parameters and the system's modes as well as a means to provide a reliable tool to determine the stability margins of the micro-grid. The configuration of the IBDG microgrid under investigation is shown in Fig. 1. The detailed control structure of the droop controlled IBDG is given in Fig. 2 [12]. The mathematical model of IBDG microgrid control loops and the complete small-signal model of the microgrid are introduced and discussed in the following subsections.

A. State Space Model of a single DG

In this subsection, the complete state space model of the IBDG with the proposed PD controller is developed based on the mathematical model introduced in [12]. As shown in Fig. 2, the three-phase output voltage and current are transformed to their dq-frame components using the angle (θ) which is obtained from the integration of the angular frequency (ω). The instantaneous active and reactive power values are computed from the voltage and current expressed in the dq-frame then are filtered through a low pass filter. The average values of the powers are fed to the power controller where the active power

droop controller sets the angular frequency of the microgrid. 1 2 The reactive power droop controller sets the reference voltage 3 and the voltage controller, in turn, sets the reference current to 4 the current controller, which is the inner most loop in this 5 control scheme. Furthermore, standard proportional integral 6 (PI) controllers are implemented in the voltage and current 7 control loops. For the sake of brevity, only the equations and 8 matrices of the modified power droop controller and its 9 reflection in the voltage source inverter (VSI) state space model 10 are derived here. The details of the voltage controller, current controller, and LC output filter state space models can be found 11 12 in [12] and are referred to in this paper using the same notation.







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Fig. 2 Detailed IBDG Microgrid Model

The power controller shares load increments between th64 IBDGs by decreasing the frequency in a way similar to the governors of synchronous generators in traditional electric power systems while adjusting the magnitude of the voltage at the PCC. The instantaneous active (P_0) and reactive (Q_0)₅ powers can be calculated in terms of the measured currents and voltages as given in (1).

$$P_0 = v_{od} i_{od} + v_{oq} i_{oq} (1)38$$

$$Q_0 = v_{oq}i_{od} - v_{od}i_{oq} \tag{133}$$

where i_{od} , i_{oq} , v_{od} and v_{oq} are the measured output currents and voltages in the dq frame, respectively. In order to reduce high fluctuations in power measurement, it is filtered through a low pass filter with a cut-off frequency (w_c). The filtered

29 powers measurement can be calculated as given in (2).

$$P = \frac{\omega_c}{S + \omega} P_0 \tag{42}$$

$$Q = \frac{\omega_c}{S + \omega_c} Q_0 \tag{2)}_{44}^{15}$$

30 The linearized state space representation of the power

controller around a specific operating point can be written usingthe first Taylor expansion as follows:

$$\Delta \dot{P} = -\omega_c \Delta P + \omega_c (I_{od} \Delta v_{od} + I_{oq} \Delta v_{oq} + V_{od} \Delta i_{od} + V_{oq} \Delta i_{oq})$$

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$$V_{od} \Delta i_{od} + V_{oq} \Delta i_{oq})$$

$$(2)$$

$$\Delta \dot{Q} = -\omega_c \Delta Q + \omega_c (I_{od} \Delta v_{oq} - I_{oq} \Delta v_{od} + V_{oq} \Delta i_{oq})$$
⁽³⁾48

33 The power controller decreases the frequency and the voltag $\frac{49}{2}$

according to the droop equations highlighted in (4).

$$\omega = \omega_n - m_P P$$

$$v_{od}^* = V_n - n_q Q$$

$$v_{oq}^* = 0$$
(4)

where ω_n and V_n are the no-load values of the frequency and voltage, respectively. n_q and m_P are the static gains of reactive and active power droop, respectively. The superscript "*" indicates the reference value.

In this paper, the static droop gain of the reactive power is replaced by a PD controller. The modified droop controller in (4) becomes:

$$\omega = \omega_n - m_P P$$

$$v_{od}^* = V_n - n_q Q - k_d \frac{dQ}{dt}$$
(5)

where k_d is the derivative gain of the reactive power droop. The proposed power controller is shown in Fig. 3.

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The linearized droop equations around the operating point in terms of the reactive and active powers are:

$$\Delta \omega = -m_P \Delta P$$

$$\Delta v_{od}^* = -n_q \Delta Q - k_d \Delta \dot{Q} \qquad (6)$$

$$\Delta v_{od}^* = 0$$

In order to convert the variables of a given inverter from the reference frame dq to the common frame DQ, the angle δ is defined as:

$$\delta = \int (\omega - \omega_{com}) dt \tag{7}$$

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Fig. 3 Power Control Loop

3 where ω_{com} is the angular frequency of the *DQ* frame of the 4 first inverter, so the differential equation for a small deviation 5 in δ is given by: 30

$$\Delta \dot{\delta} = \Delta \omega - \Delta \omega_{com} \tag{8)31}$$

6 The small-signal state-space representation of the modified 2 7 power controller can be written combining the equations given 3 in (3), (6), and (8) as follows:

$$\begin{bmatrix} \Delta \dot{\delta} \\ \delta \end{bmatrix} = \begin{bmatrix} \Delta \delta \\ \delta \end{bmatrix} = \begin{bmatrix} \Delta i_{ldq} \\ \delta i_{ldq} \end{bmatrix}$$

$$\begin{array}{c} \overline{\Delta P} \\ \Delta P \\ \Delta Q \end{array} = A_P \begin{bmatrix} \overline{\Delta P} \\ \Delta Q \end{bmatrix} + B_P \begin{bmatrix} \Delta w_{odq} \\ \Delta v_{odq} \\ \Delta i_{odg} \end{bmatrix} + B_{Pwcom} [\Delta \omega_{com}] \qquad (9) \end{array}$$

$$\begin{bmatrix} \Delta w \\ \Delta v_{odq}^* \end{bmatrix} = \begin{bmatrix} C_{Pw} \\ C_{Pv} \end{bmatrix} \begin{bmatrix} \Delta \delta \\ \Delta P \\ \Delta Q \end{bmatrix} + \begin{bmatrix} D_{Pw} \\ D_{Pv} \end{bmatrix} \begin{bmatrix} \Delta i_{ldq} \\ \Delta v_{odq} \\ \Delta i_{odq} \end{bmatrix}$$
(10)₃₆

$$\Theta \quad \text{where, } A_P = \begin{bmatrix} 0 & -m_P & 0 \\ 0 & -\omega_c & 0 \\ 0 & 0 & -\omega_c \end{bmatrix}, \Delta i_{ldq} = \begin{bmatrix} \Delta i_{ld} \\ \Delta i_{lq} \end{bmatrix}, \qquad 38 \\ 39 \\ 40 \end{bmatrix}$$

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$$\Delta v_{odq} = \begin{bmatrix} \Delta v_{od} \\ \Delta v_{oq} \end{bmatrix}$$
, $\Delta i_{odq} = \begin{bmatrix} \Delta i_{od} \\ \Delta i_{oq} \end{bmatrix}$, $\Delta v_{odq}^* = \begin{bmatrix} \Delta v_{od}^* \\ \Delta v_{oq}^* \end{bmatrix}$

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$$B_P = \begin{bmatrix} 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & \omega_c I_{od} & \omega_c I_{oq} & \omega_c V_{od} & \omega_c V_{oq} \\ 0 & 0 & -\omega_c I_{oq} & \omega_c I_{od} & \omega_c V_{oq} & -\omega_c V_{od} \end{bmatrix}$$

12
$$C_{Pv} = \begin{bmatrix} 0 & 0 & -n_q + \omega_c k_d \\ 0 & 0 & 0 \end{bmatrix}, C_{Pw} = \begin{bmatrix} 0 & -m_P & 0 \end{bmatrix},$$

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$$D_{PW} = \begin{bmatrix} 0 & 0 & 0 & 0 & 0 \end{bmatrix}, B_{Pwcom}^{T} = \begin{bmatrix} -1 & 0 & 0 \end{bmatrix},$$

14 $D_{Pv} = \begin{bmatrix} 0 & 0 & k_d \omega_c I_{oq} & -k_d \omega_c I_{od} & -k_d \omega_c V_{oq} & k_d \omega_c V_{od} \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix}$

15 Using the transformations given in (11) and (12), the output 8 16 variables (Δw and Δv_{odq}^*) and input variable ($\Delta \omega_{com}$) of the 9 17 inverter can be transferred from the dq frame to the common 0 18 DQ frame as follows: 51

$$\Delta i_{oDQ} = T_s \,\Delta i_{odq} + T_C \,\Delta\delta \tag{11}^{52}$$

$$\Delta v_{bdq} = T_s^{-1} \Delta v_{bDQ} + T_v \Delta \delta \tag{12}$$

19 where, δ_0 is the steady-state power angle, and

$$S_{s} = \begin{bmatrix} \cos(\delta_{0}) & -\sin(\delta_{0}) \\ \sin(\delta_{0}) & \cos(\delta_{0}) \end{bmatrix}$$
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$$\begin{bmatrix} -I_{od}si n(\delta_0) - I_{oq} cos(\delta_0) \end{bmatrix}$$
⁵⁴
₅₅

$$T_c = \begin{bmatrix} I_{od} cos(\delta_0) - I_{oq} sin(\delta_0) \end{bmatrix}$$

$$\left[-V_{bD} \sin n(\delta_0) + V_{bO} \cos(\delta_0)\right]$$
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$$T_{v} = \begin{vmatrix} V_{bD} \cos(\delta_{0}) + V_{bQ} \cos(\delta_{0}) \\ -V_{bD} \cos(\delta_{0}) - V_{bQ} \sin(\delta_{0}) \end{vmatrix}$$

20 The complete model of the inverter is obtained through the

combination of the state space models of each subsystem. The
overall state-space model of the inverter consists of 13 states,
and it can be written as follows:

$$\Delta \dot{x}_{invi} = A_{invi} \Delta x_{invi} + B_{invi} \Delta v_{bDQi} + B_{iwcom} \Delta \omega_{com}$$
(13)

$$\begin{bmatrix} \Delta w_i \\ \Delta i_{oDQi} \end{bmatrix} = \begin{bmatrix} C_{invwi} \\ C_{invci} \end{bmatrix} \Delta x_{invi}$$
(14)

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where $\Delta x_{invi} = [\Delta \hat{\delta}_i \ \Delta P_i \ \Delta Q_i \ \Delta \phi_{dqi} \ \Delta \gamma_{dqi} \ \Delta i_{ldqi} \ \Delta v_{odqi} \ \Delta i_{odq}]^T$, $\dot{\gamma}_{dq} = i_{ldq}^* - i_{ldq}$ and $\dot{\phi}_{dq} = V_{odq}^* - V_{odq}$.

Finally, the small-signal state space model of the three DGs in the microgrid under investigation can be derived based on the state space model of each individual inverter given in (13) and (14) as follows:

$$\Delta \dot{x}_{INV} = A_{INV} \Delta x_{INV} + B_{INV} \Delta v_{bDQ} \tag{15}$$

$$\Delta i_{oDQ} = C_{INVc} \,\Delta x_{INV} \tag{16}$$

where $\Delta x_{INV} = [\Delta x_{inv1} \quad \Delta x_{inv2} \quad \Delta x_{inv3}]^T$, A_{INV} , B_{INV} and C_{INVc} are the state space representation of the combined inverters.

B. Load and Network Model

The network and load small-signal model of the microgrid shown in Fig. 1 can be described as follows:

$$\dot{\Delta i}_{lineDQ} = A_{NET} \Delta i_{lineDQ} + B_{1NET} \Delta v_{bDQ} + B_{2NET} \Delta w \qquad (17)$$

$$\Delta i_{loadDQ} = A_{Load} \Delta i_{LoadDQ} + B_{1Load} \Delta v_{bDQ} +$$
(18)

 $B_{2Load}\Delta w$ where $\Delta i_{lineDQ} = [\Delta i_{lineDQ1} \quad \Delta i_{lineDQ2}]$ is the line current in the DQ frame. $\Delta v_{bDQ} = [\Delta v_{bDQ1} \quad \Delta v_{bDQ2} \quad \Delta v_{bDQ3}]$ is the bus voltage in the DQ frame. $\Delta w = \Delta w_{com}, \quad \Delta i_{loadDQ} = [\Delta i_{loadDQ1} \quad \Delta i_{loadDQ2}]. A_{NET}, \quad B_{1NET}, \quad B_{2NET}, A_{Load}, B_{1Load},$ and B_{2Load} are the network mode matrices.

C. Overall Microgrid Model:

In order to write a complete microgrid model, the input variable (Δv_{bDQ}) in the network and load models given in (17) and (18) can be obtained by assuming the existence of a large virtual resistor r_N connected between each node and ground. Thus, the small-signal model of a node can be given by:

$$\Delta v_{bDQ} = R_N \{ M_{INV} \Delta i_{oDQ} + M_{Load} \Delta i_{loadDQ} + M_{Net} \Delta i_{ineDQ} \}$$
(19)

where the matrices M_{INV} , M_{Load} , and M_{Net} map the DGs connection points, the loads connection points, and the lines connection points to the nodes, respectively. For the microgrid configuration shown in Fig. 1, the complete microgrid small-signal state-space model can be obtained by combining the systems given in (15)-(19).

$$\begin{bmatrix} \Delta x_{INV} \\ \Delta i_{lineDQ} \\ \Delta i_{loadDQ} \end{bmatrix} = A_{MG} \begin{bmatrix} \Delta x_{INV} \\ \Delta i_{lineDQ} \\ \Delta i_{loadDQ} \end{bmatrix}$$
(20)

The details of the matrices M_{INV} , M_{Load} , M_{Net} and the microgrid's state matrix A_{MG} are omitted here because of space limitations and can be constructed as given in [12]. The main modification between the model represented here and the model given in [12], is in the power control loop which directly affects the matrices A_{invi} , C_{invwi} and consequently. A_{MG} .

$$1 \qquad A_{invi} = \begin{bmatrix} A_{Pi} & 0 & 0 & B_{Pi} \\ B_{v1i}C_{Pvi} & 0 & 0 & B_{v2i} + B_{v1i}D_{Pvi} \\ B_{c1i}D_{v1i}C_{Pvi} & B_{c1i}C_{vi} & 0 & B_{c1i}D_{v2i} + B_{c2i} + B_{c1i}D_{v1i}D_{Pvi} \\ B_{LCL1i}D_{c1i}D_{v1i}C_{Pvi} + B_{LCL3i}C_{Pwi} & B_{LCL1i}D_{c1i}C_{vi} & A_{LCLi} + B_{LCL1i}D_{c1i}D_{v2i} + B_{LCL1i}D_{c2i} \\ + B_{LCL2i}[T_{vi}^{-1} & 0 & 0] & + B_{LCL3i}D_{Pwi} + B_{LCL1i}D_{c1i}D_{v1i}D_{Pvi} \\ \end{bmatrix}_{13\times13}$$

$$2 \qquad B_{invi} = \begin{bmatrix} 0 \\ 0 \\ 0 \\ B_{LC2i}T_{Si}^{-1} \end{bmatrix}_{13\times2} B_{iwcom} = \begin{bmatrix} B_{Pwcom} \\ 0 \\ 0 \\ 0 \end{bmatrix}_{1\times13}, C_{invvi} = \begin{cases} [C_{Pwi} & 0 & 0 & D_{Pwi}]_{1\times13} & i = 1 \\ [0 & 0 & 0 & D_{Pwi}]_{1\times13} & i \neq 1 \end{cases}, c_{invci} =$$

$$3 \qquad [[T_{C} & 0 & 0] & 0 & 0 & [0 & 0 & T_{S}]]_{2\times13}, \Delta v_{bDQ} = [\Delta v_{bDQ1} \quad \Delta v_{bDQ2} \quad \Delta v_{bDQ3}]$$

$$4 \qquad 5 \qquad A_{INV} = \begin{bmatrix} A_{inv1} + B_{1wcom}C_{invw1} & 0 & 0 \\ B_{2wcom}C_{invv1} & A_{inv2} & 0 \\ B_{3wcom}C_{invv1} & 0 & A_{inv3} \end{bmatrix}_{39\times39}, B_{INV} = \begin{bmatrix} B_{inv1} \\ B_{inv2} \\ B_{inv3} \end{bmatrix}_{39\times2}$$

$$6 \qquad C_{INVc} = \begin{bmatrix} C_{invc1} & 0 & 0 \\ 0 & C_{invc2} & 0 \\ 0 & 0 & C_{invc3} \end{bmatrix}_{6\times39}, C_{INVw} = [[C_{Pw1} & 0 & 0 & D_{Pw1}] \quad [0 & 0 & 0 & D_{Pw2}] \quad [0 & 0 & 0 & D_{Pw3}]]_{1\times39},$$

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III. STABILITY DOMAIN ANALYSIS

8 To investigate the effect of the active and reactive powe#4 9 droop controllers' parameters on the stability of the microgrid45 10 the benchmark IBDG microgrid in [12] is considered for 6 11 comparative purposes. The main parameters of this microgrid7 12 are given in Table I. The initial steady-state operating point i 48 13 obtained and then the microgrid stability analysis is performed9 14 on the small-signal model given in (20). 50

15 In the literature on droop-controlled microgrids, stability 16 analysis is determined based solely on the active power droop 17 gain (m_p) . As shown in [12], increasing the range of the active power droop gain leads to a reduction in the settling time of the 5^3 18 19 microgrid and increases the ability of the microgrid to handl ℓ^4 20 different loading conditions. Moreover, increasing the active5 21 power droop gain above a specific maximum value leads $t\delta^6$ instability of the microgrid, as can be observed from the? 22 eigenvalue analysis. In this section, the effect of changing the δ^8 23 reactive power droop gain (n_q) on the stability domain $i\delta^9$ 24 60 25 investigated. 61

26 A. Impact of Reactive Power Droop gain on Stability

27 In this subsection, the correlation between both active an**6**3 28 reactive droop gains is investigated by observing the effect of 4 29 changing the reactive power droop gain (n_a) on the operating 5 30 range of the active power droop gain. The eigenvalues of the6 31 microgrid small-signal model system in (20) are plotted in 7 Fig. 4. The eigenvalues of the microgrid can be divided int68 32 33 three clusters, out of which cluster 1 is the most critical as i69 34 contains the dominant poles of the microgrid. 70

The sensitivity analysis in [12] shows that the dominant 35 36 modes of the microgrid are highly sensitive to the active droop₂ gain (m_v) while exhibiting a much lesser sensitivity to the₃ 37 reactive droop gain. The variations of the dominant $pole_{\overline{A}}$ 38 location with m_p are plotted in Fig. 5. Increasing the active 5539 droop gain pushes a pair of the dominant poles towards the 40 imaginary axis with marginal stability occurring at $m_p = \frac{1}{77}$ 41 $1.9e^{-4}$. In Fig. 6, the loci of dominant poles in cluster 1 are 42 plotted as n_q is increased while fixing the active power droop_Q 43

at $m_p = 0.9e^{-4}$. Increasing n_q first pushes the dominant pair of poles towards the imaginary axis then moves them back away from it. Yet, this increase in the reactive droop gain drives a second pair of complex poles towards the imaginary axis, eventually making them the new dominant pair. This observation means that the maximum active power droop gain that preserves stability can be significantly increased by adjusting the reactive power droop gain (n_q) .

B. Proposed Stability Domain Chart for Inverter Based Micro-grids

As indicated in the IEEE Std. 1547, while planning microgrids, it is essential to determine the acceptable stability limits [1]. For inverter-based microgrids, a detailed trainset model of the IBDG and its control is essential for assessing the stability of the microgrid [1]. As seen in the previous subsection, both active and reactive power droop gains play a significant role in identifying the microgrid stability. Thus, in this paper, a stability domain chart is proposed to not only assess the system stability but will also be used to define the microgrid operable stable region.

Fig. 7 clearly shows the correlation between the active and reactive droop gains, where the maximum allowable m_p that maintains the microgrid stability is plotted against n_a . Fig. 7 represents the stability domain chart, where the region encompassed within the blue line represents all possible microgrid stable operating points. Increasing n_q results in an initial decrease in the maximum active droop gain (m_{pmax}) followed by a significant increase in its value that eventually leads to instability at any value of m_p when n_q exceeds 5.9e⁻³. This observation is illustrated in Fig. 8 by plotting the loci of the dominant poles as m_p varies from $1.9e^{-6}$ to $3e^{-4}$ at n_q of 1.3e-3. Increasing n_q from 1.3e-3 to 4e-3, approximately by 200%, results in an increase in the maximum active power droop gain by 50 %, i.e. m_{pmax} increased from 1.9e-4 to 3e-4. Using the proposed stability domain chart, it can be seen that the reactive power droop controller can be adjusted to improve

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the stability margin of the microgrid. Furthermore, the stability domain chart can be used in assessing the various DG interface control schemes for micro-grid operation, as will be seen in the next subsection.

TABLE I Microgrid Parameters				
Parameter	Value	Parameter	Value	
R _f	0.1	R _{load2}	25	
L_{f}	1.35e-3	L_{load2}	0.1e-1	
C_{f}	50e-6	m_p	9.5e-5	
R_c	0.03	n_q	1.3e-3	
L_c	0.35e-3	k_d	0	
R_{line1}	0.23	k _{pc}	10.5	
L_{line1}	0.35e-3	k_{ic}	16e3	
R_{line2}	0.35	k _{pv}	0.05	11
L_{ine2}	1.85e-3	k _{iv}	390	12
R_{load1}	25	W _c	31.41	14
L_{load1}	0.1e-1	F	0.75	







C. Stability Domain Chart of PD Reactive Power Controller

The stability domain chart will be utilized to assess the impact of the PD reactive power controller on the microgrid stability as well as to compare its performance with the conventional reactive power droop. In order to further improve the stability margin, the reactive power droop gain is replaced by a PD controller. The new domain of stability is investigated based on the eigenvalue analysis of the modified small-signal model derived in Section II.

26 The first tuning parameter available to increase the domain 27 of stability is the derivative gain (k_d) of the PD controller. The effect of varying the derivative gain on the dominant poles 28 locations at $n_q = 1.3e^{-3}$ is shown in Fig. 9. The maximum active 29 power droop gain increases from $1.9e^{-4}$ to $5.47e^{-4}$ and 30

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 $8.56e^{-4}$ when the derivative gain is increased from 0 to 1 $0.625e^{-4}$ and $1.25e^{-4}$, respectively. 2

3 The second tuning parameter is the proportional gain (n_a) of the PD controller. The effect of changing the proportional gain 4 on the dominant poles locations at $k_d = 0.625e^{-4}$ is shown in 5 Fig. 10. The results of the analysis summarized in Table II 6 7 clearly show that the maximum active power droop can be significantly increased using the PD control in the reactive 8 9 power droop control. The stability domain is increased by 10 58.9% and 142.6% when the derivative gain is doubled at the same reactive droop gain ($n_q = 5.9e^{-3}$). Fig. 11 presents the 25 11 domain of stability chart for the conventional and PD reactive 12 power droop. As can be seen, the domain of stability is a_1^{26} 13 effective tool in assessing and quantifying the effect of DG'_{28} 14 15 interface control on microgrid stability.



Reactive power droop gain (n_q)	Derivative reactive power gain (k_d)	Maximum active power droop gain (m_{pmax})
	0	$2.33e^{-4}$
0	$0.625e^{-4}$	$5.6e^{-4}$
	$1.25e^{-4}$	$8.5e^{-4}$
	0	$1.9e^{-4}$
1.3e ⁻³	$0.625e^{-4}$	$5.47e^{-4}$
	$1.25e^{-4}$	$8.56e^{-4}$
	0	$3.73e^{-4}$
$5.9e^{-3}$	$0.625e^{-4}$	5.93e ⁻⁴
	$1.25e^{-4}$	9.05e ⁻⁴

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D. Stability Margin Chart of Reactive Power Controller

The stability domain provides the region within which the microgrid is stable, but does not take into account the microgrid transient performance. In order to achieve both transient and steady-state stability, a stability margin region is identified within the stability domain chart. The stability margin discussed in this paper is defined as the angle between the imaginary axis and the dominant eigenvalue [18]. The larger angle indicates the higher damping of the oscillatory modes, which directly affects the transient performance of the microgrid. Fig. 12 presents the domain of stability chart amended with the stability margin operating regions for the reactive power droop controller (n_a) . The operating range of the microgrid decreases with the increase in the stability margin. However, increasing the stability margin directly improves the transient performance. As can be seen, there is a value for n_a beyond which the stability margin significantly decreases. Similarly, the stability domain chart amended with the stability margin operating region for the PD reactive power controller is shown in Fig. 13. The PD controller provides a larger stability margin region and thus providing the microgrid operator with a wider range for the microgrid droop gains.



Fig. 12 Stability margin chart without a PD controller.



IV. SIMULATION RESULTS

5 The domain of stability not only provides all possible stable 6 operating conditions, but it is also important to maintain 7 acceptable transient performance by selecting the operating 8 stability margin. The simulation results are used to validate the 9 eigenvalue stability analysis, assess the microgrid performance, 10 and analyze the stability domain benefits. In the first subsection, 11 verification of the stability domain chart in representing the microgrid stable operating points is conducted. In the next 12 13 subsection, the stability margin region identified within th58 stability domain chart, in the previous section, is tested. The 14 15 impact of load change, line disconnection, and DG 16 disconnection are studied in subsections C, D and E, 17 respectively.

18 A. Stability Domain Chart Verification

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19 As discussed in subsection III.C, the stability domain 20 (maximum allowable active power droop (m_{pmax})) can be 21 increased by increasing either the reactive power gain (n_a) of $\frac{1}{2}$ 22 the derivative gain of the reactive power (k_d) . In order to test the stability domain chart plotted in Fig. 11, point A is selected. 23 24 Point A is selected such that it is outside the stability domain 25 for $k_d = 0$ and 0.625e-4 and within the stability domain for the 26 case where $k_d = 1.25e-4$. The active power and frequency for 27 the three DGs are shown in Fig. 14 and Fig. 15 for $k_d = 1.25e$ -28 4. As shown in these figures, at t = 0 sec the load is connected 29 and the simulation is started, the response of the microgrid 30 includes high oscillations during the transient period, and the 31 system goes to steady-state after 1.5 seconds. The reason behind 32 these oscillations and the higher settling time is the close 33 vicinity of operating point A to the stability domain boundary. The active power responses of the three DGs when $k_d = 0$ and k_d^3 34 35 0.625e-4 have been tested and were shown to result in unstable 5 36 operation.

37 B. Stability Margin Region Validation

38 In this subsection, the stability margin regions shown in Fig. 39 12 and Fig. 13 are tested. Point B is selected to identically set all the DGs active and reactive droop gains at $9.5e^{-5}$ and $1.3e^{-3}$, 40 41 respectively. By referring to these figures, it can be seen that 42 point B is located within the stability margin region of 10 43 degrees and 15 degrees for conventional and PD reactive power 44 droop controllers, respectively. The results at point B of the 45 conventional reactive power droop controller given in (4) and the proposed PD reactive power droop controller given in (5) are compared. The results for the DG active and reactive powers as well as microgrid frequency are presented in Fig. 16, Fig. 17 and Fig. 18. For the same droop gain values, the proposed PD reactive power controller achieves a better transient response in terms of the settling time and overshoot which coincides with the results presented in the domain of stability chart.

Thus, the proposed domain of stability chart is a useful tool that can aid in micro-grid operators in: 1) selecting the droop gains to achieve desirable microgrid transient and steady-state performance, and 2) assessing and comparing different microgrid controllers.



Fig. 14 Active power of the three DGs at the nearest critically stable operating



Fig. 16 Active power sharing response of the three DGs of the conventional versus PD reactive droop controllers.



In this case study, the effect of different loading conditions 8 9 on the microgrid stability and, more specifically, on the domain 10 of stability is analyzed. The total capacity of the three DGs 11 under study is 30 kW. In order to examine the effect of different loading conditions on the microgrid stability, three different 12 13 loading conditions corresponding to 40%, 80% and 100% of 14 rated loading are tested. The stability domains of the three loading conditions are shown in Fig. 19. It can be seen that the 15 stability domains at 40%, 80% and 100 % are identical, which 16 17 indicated that different loading levels have no effect on the 18 stability domain. To further validate the results, time domain simulation was conducted using SIMULINK/MATLAB considering load changes. The active and reactive powers of the 4319 20 21 three DGs are shown in Fig. 20 and Fig. 21, respectively.

22 To validate the effect of different loading conditions on the 23 microgrid stability, the microgrid system is operated at 40 % 24 loading for 0.5 sec, and the loading is increased to 80% for 0.5 25 sec and then it is decreased to 40 % for 0.5 sec. The active and 26 reactive droop gains for all DGs are identical and are set to 27 $0.95e^{-4}$ and $1.3e^{-3}$, respectively. The results show that the 28 droop controller is capable of maintaining the microgrid 29 stability with load changes. Although the microgrid frequency is slightly affected by the loading condition, as shown in Fig₁₄ 30 31 22, the microgrid frequency is within the normal permissible 5 levels. Lastly, the transient performance of the proposed PD_{46} 32 33 reactive droop controller is compared with the reactive gain controller in Fig. 20 and Fig. 21. The proposed PD reactive⁴⁷ 34 droop controller improves the transient performance⁴⁸ 35 49 36 significantly under all loading conditions. 50





D. Impact of Line disconnection on Domain of Stability

During normal operation of the microgrid, one of the microgrid lines may be disconnected due to any abnormal conditions which could impact the microgrid stability. In order to validate the robustness of the microgrid towards line disconnection, T.L. 2 shown in Fig. 1 is disconnected, which results in two separate microgrids. For the same values of the active and reactive droops, the proposed PD reactive controller performance is compared with the reactive droop gain. It is

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1 assumed that the microgrid is operating at normal conditions 2 and T.L. 2 is disconnected at t = 2 sec. The active and reactive 3 power-sharing of the DGs during the normal operation and 4 sudden disconnection of T.L. 2 are shown in Fig. 23 and Fig. 5 24.

6 The active and reactive droop gains for all DGs are identical and are set to $1.8e^{-4}$ and $1.3e^{-3}$ which are labeled as point C in 7 8 Fig. 25. The results confirm the capability of the proposed PD 9 reactive controller to maintain microgrid stability while the 10 reactive droop gain fails to maintain stable operation. The 11 reactive droop gain has a smaller domain of stability with 12 respect to the PD reactive controller. Fig. 25 shows the domain 13 of stability for the PD reactive droop with and without $lin\bar{\vartheta}0$ disconnection at $k_d = 1.5e^{-4}$. To maintain stable micro-grid 14 operation with and without line disconnection, the overall 15 domain of stability represents the intersection region between³² 16 the two domains of stability shown in Fig. 25. This are area i^{33} 17 labeled as PQRS and as can be seen, point C lies within this⁴ 18 area, and thus stable operation is achieved with the PD reactive $\frac{3}{2}$ 19 power controller before and after line disconnection. The result36 20 show that increasing the domain of stability using the PD^{7} 21 reactive controller can enhance the overall micro-grid stability38 22 39 23 during abnormal conditions. 40





E. Impact of DG disconnection on Domain of Stability

To further validate the application of the domain of stability, the effect of disconnecting one of the DGs on the microgrid stability is analyzed. The DGs share the active and reactive power during the normal operation and a sudden disconnection of DG3 occurs at t = 2 sec. Fig. 26 and Fig. 27 present the DG active and reactive power output considering both reactive power droop and PD reactive controller, respectively. The active and reactive droop gains for all DGs are identical and are set to $1.82e^{-4}$ and $1.3e^{-3}$, respectively which is represented by point D in Fig. 28. Similarly, Fig. 28 presents the domain of stability for the PD reactive power controller considering both normal and DG3 disconnection at $k_d = 1.5e^{-4}$. The overall domain of stability is the intersection region labelled as WXYZ in Fig. 28. The results show that the proposed PD reactive droop controller has the ability to maintain the microgrid stability in comparison to the reactive power controller which fails to maintain stable operation. As can be seen, point D lies within the domain of stability region WXYZ resulting in a stable operating point. The aforementioned case studies highlight the capability of the PD reactive controller in enhancing the microgrid stability and increasing the domain of stability allowing it to maintain stable operation during normal and abnormal conditions.



Fig. 26 Active power-sharing in the case of DG3 disconnected.



6 This paper proposes a domain of stability chart for defining 7 the microgrid operating region, considering both transient and steady-state operation. The stability domain chart is developed 8 9 using eigenvalue analysis and is applied to the conventional as 10 well as PD reactive power controller. Investigating the 11 maximum active power droop while fixing the reactive power 12 droop gain results in an unnecessary limitation on enhancing 13 the stability margin, and consequently, the domain of stability. 14 Using the proposed chart, the results show that for both 15 controllers, higher stability margins can be achieved by proper selection of both the active and reactive power droop gains. 16

Furthermore, the maximum active droop gains presented in 17 18 previous literature can be significantly increased by optimally 19 selecting the reactive droop gain. Both active and reactive 20 power droop can have a significant effect in defining the 21 microgrid domain of stability. The domain of stability chart also 22 validates that a derivative controller in the reactive power 23 controller can sufficiently enhance the microgrid transient and 24 steady-state performance. The domain of stability chart can 25 serve as a useful tool for identifying the micro-grid operational 26 boundaries as well as for assessing and comparing micro-grid 27 droop based control schemes.

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